

Low-Noise Voltage Preamplifier

SR560 — DC to 1 MHz voltage preamplifier



SR560 Low-Noise Preamplifier

- **4 nV/ $\sqrt{\text{Hz}}$ input noise**
- **1 MHz bandwidth**
- **Variable gain from 1 to 50,000**
- **AC or DC coupled**
- **Two configurable signal filters**
- **Differential and single-ended inputs**
- **Line or battery operation**
- **RS-232 interface**

• **SR560 ... \$2195 (U.S. list)**

The SR560 is a high-performance, low-noise preamplifier that is ideal for a wide variety of applications including low-temperature measurements, optical detection, and audio engineering.

Inputs

The SR560 has a differential front-end with 4 nV/ $\sqrt{\text{Hz}}$ input noise and an input impedance of 100 M Ω . Complete noise figure contours are shown in the SR560 Tech Note at the end of this section. The SR560's inputs are fully floating (BNC shields are not connected to chassis ground). Both the amplifier ground and the chassis ground are available on the rear panel for flexibility in grounding the instrument. Input offset nulling is accomplished by a front-panel potentiometer, accessible with a small screwdriver.

In addition to the signal inputs, a rear-panel TTL blanking input lets you quickly turn off and on the instrument's gain. This is useful in preventing front-end overloading. The gain turns off 5 μs after the TTL level goes high, and back on again within 10 μs after the TTL signal goes low.

Outputs

Two insulated output BNC connectors provide 600 Ω and 50 Ω outputs. Both are capable of driving 10 V_{pp} into their respective loads. Two rear-panel power supply outputs provide up to 200 mA of ± 12 VDC referenced to the amplifier

Input

Inputs	AC or DC coupled, single-ended or differential
Input impedance	100 M Ω + 25 pF
Maximum input	3 Vpp
CMRR	100 dB from DC to 1 kHz (Decreases by 6 dB/octave from 1 kHz to 1 MHz)
Noise	4 nV/ $\sqrt{\text{Hz}}$ at 1 kHz
Gain	1 to 50,000 in 1–2–5 sequence Vernier gain in 0.5 % steps
Gain stability	200 ppm/ $^{\circ}\text{C}$
Frequency response	± 0.5 dB to 1 MHz, ± 0.3 dB to 300 kHz (gains up to 1000)

Filters

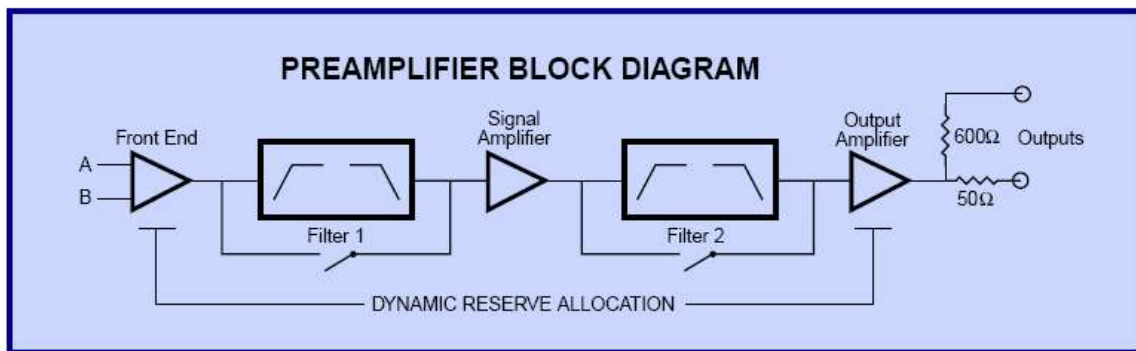
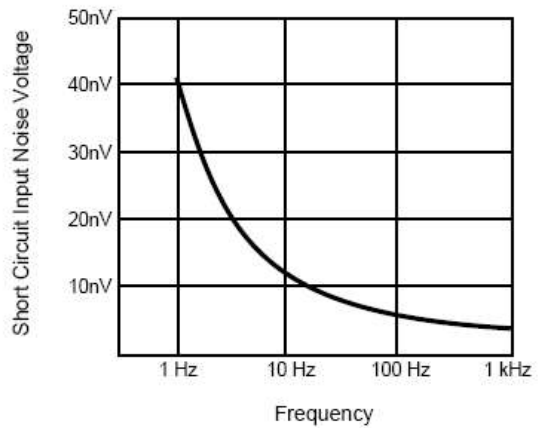
Signal filters	2 configurable (low-pass or high-pass) 6 dB/oct rolloff filters. –3 dB points are settable in a 1–3–10 sequence from 0.03 Hz to 1 MHz.
Gain allocation	High Dynamic Reserve—Gain is increased after the signal filters to prevent overloading. Low Noise—Gain is increased before the filters to improve noise figure.

Output

Maximum output	10 Vpp into 50 Ω and 600 Ω
Filter reset	Long time constant filters may be reset with front-panel button.
DC drift	5 $\mu\text{V}/^{\circ}\text{C}$ referred to input (DC coupled)
Distortion	0.01 % at 1 kHz
Rear panel	± 12 VDC @ 200 mA referenced to amplifier ground

General

External gating	TTL input sets gain to zero
Interfaces	RS-232, 9600 baud, receive only
Power	100/120/220/240 VAC, 6 W charged, 30 W while charging. Internal batteries provide 15 hours of operation between charges. Batteries are charged while connected to the line.
Dimensions	8.3" \times 3.5" \times 13.0" (WHD)
Weight	15 lbs. (batteries installed)
Warranty	One year parts and labor on defects in materials and workmanship



Noise Figure Contours

Noise figure contours are often provided with amplifier specifications, but many users are unclear on what they signify. The noise figure of an amplifier is the ratio (usually expressed in dB) of the equivalent input noise of the amplifier, at a given frequency, to the thermal noise of a source with a given source impedance. Since the equivalent input noise of an amplifier is simply the output noise divided by the gain, this can be expressed as:

$$NF = 20 \log (\text{Output-Noise/Gain})/(\text{Source Thermal Noise})$$

If the amplifier were noiseless, all the output noise would be due to the source thermal noise, and the noise figure would be 0 dB. To the extent that the amplifier adds some of its own noise, the Output-Noise/Gain will be bigger than the source thermal noise, and the noise figure will be non-zero. Thus, the noise figure is an indication of how much of the output noise is contributed by the amplifier.

Since amplifier noise and thermal noise are functions of both frequency and source impedance, noise figures are often

plotted as contours on a graph of source resistance versus frequency. At a given frequency, the noise figure is large for very low source resistance since the thermal noise of the source is small, and any amplifier noise leads to a large noise figure. As the source resistance increases, the noise figure becomes better until the current noise of the amplifier passing through the high source resistance once again degrades the noise figure.

At a constant source resistance, the noise figure is poor at low frequencies because of the 1/f noise of the amplifier. As the frequency is increased, 1/f noise decreases and the noise figure improves. Eventually, for all transistors, the current noise increases as a function of frequency, and the noise figure begins to increase again.

Does this mean that you should always increase the resistance of your source to reach the region of minimum noise figure? Certainly not. By doing that you would only be making your source noisier to improve the noise figure. Remember, the goal is to reduce total noise, not to minimize the noise figure.

SR560 Noise Figure

